

## High Output Voltage Based Multiphase Step-Up DC-DC Converter Topology with voltage doubler rectifiers

Amir Mahmood Soomro<sup>\*1</sup>, Xu lingyu<sup>2</sup>, Shahnawaz Farhan Khahro<sup>3</sup>, Liao Xiaozhong<sup>4</sup>

<sup>1,2,3,4</sup>School of Automation, Beijing Institute of Technology, Beijing 100081, P. R. China

<sup>1</sup>Department of Electrical Engineering, Mehran University of Engineering and Technology, Jamshoro, Sindh, Pakistan

\*Corresponding author, e-mail: aamir.m.soomro@gmail.com<sup>1</sup>, bitxly@gmail.com<sup>2</sup>, shahnawazfarhan@gmail.com<sup>3</sup>, liaoxiaozhong@bit.edu.cn<sup>4</sup>

### Abstract

*High Output Voltage Based Multiphase Step-Up DC-DC Converter topology with voltage doubler rectifiers is presented in this paper. High output voltage is obtained due to the series combination of voltage doubler rectifiers on the secondary side of high frequency transformers. This topology is useful in the application where the output voltage is greater than the input. The two loop control strategy has been developed in order to analyze the stable and effective working of the converter topology. Therefore the working mode analysis of the converter topology has been described in detail. The multiphase step-up DC-DC converter topology is first simulated on MATLAB and then a prototype has been designed in order to verify the simulation and experimental results. Finally the simulation and experimental results are found to be satisfactory.*

**Keywords:** DC-DC converter, High frequency transformers (HFTs), Multiphase, Voltage doubler rectifiers (VDRs).

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## 1. Introduction

Nowadays multiphase step-up DC-DC converter plays an important role in the field of power electronics and control. Therefore different topologies have been proposed [1-16] for high power and high voltage applications. Some of those topologies are of current fed type [1-5] and few of them are of voltage fed type [6-10]. The first multiphase DC-DC converter has a limitation of switching frequency because soft commutation was not introduced in it [6]. Active clamp technique has been introduced to compensate the duty cycle loss, reduces circulating current and suppresses surge voltage [9]. V6 converter becomes beneficial for high power applications but at the cost of increased number of switches [10]. Multiphase DC-DC converter has number of benefits over single phase DC-DC converter. The primary advantage is the reduced current stress on the devices due to the fact that currents are shared among the phases. This paper presents the multiphase step-up DC-DC converter topology with voltage doubler rectifier. A comprehensive multiphase DC-DC converter topology is designed to increase output voltage; therefore voltage doubler rectifiers in series configuration have been connected at the secondary side of high frequency transformers. As compared to the input DC voltage, High output DC voltage is obtained and is considered to be the main advantage of the proposed topology. Furthermore, the efficiency is improved due to the reduced number of switches and the voltage step-up characteristic. The output filter volume reduces along with a high switching frequency.

## 2. Proposed Multiphase Step-up DC-DC Converter Topology

### 2.1 Circuit Description

The multiphase step-up DC-DC converter topology with voltage doubler rectifiers (VDRs) is represented in Figure. 1 The topology consists of dc input voltage source, three input inductors and three MOSFETs on the primary side of high frequency transformers (HFTs), whereas the voltage doubler rectifiers (VDRs) in series combination are connected on the secondary side of high frequency transformers (HFTs). The main purpose of connecting voltage

doubler rectifiers (VDRs) in series combination at the secondary side of high frequency transformers (HFTs) is to obtain high output voltage. The main characteristics of multiphase step-up DC-DC converter topology are: Same reference is shared by only three switches, featuring simplicity to the drive and a control circuits; Amount of voltage at the switches is decreased because of the employed transformer.

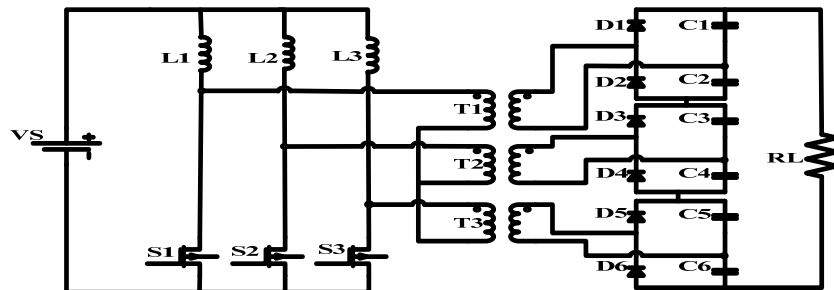


Figure 1. Multiphase Step-up DC-DC Converter

## 2.2 Control Strategy

Control strategy of multiphase step-up DC-DC converter topology with voltage doubler rectifiers (VDRs) is shown in Figure 2. The control strategy consisting of two loops, one of them is called output voltage loop and the other one is said to be input current loop. Output voltage loop needs to sense the output voltage  $V_o$ , then added/subtracted with the reference value  $V_{ref}$  and generates the error signal. The error signal is given as an input signal to the  $PI_1$ -controller. The output of the  $PI_1$ -controller acts as the reference current  $I_{ref}$  for the input current loop. Similarly the input current  $I_{in}$  is sensed by input current loop and is added/subtracted with the reference current  $I_{ref}$  in order to generate the error signal as an input to the  $PI_2$ -controller. The output of the  $PI_2$ -controller is given to PWMs in order to generate the driving signals for the gates of three switches.

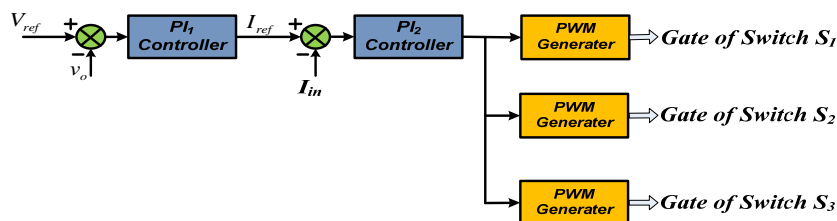


Figure 2. Control Strategy

## 2.3 Working Mode Analysis

The proposed multiphase step-up DC-DC converter topology has three different operation areas. Each one varies from other by the number of switches ON at the same time. Duty cycle from 0 to 33.33% is the first area and is denoted by  $A_1$ . Duty cycle from 33.33% to 66.67% is the second area and is represented by  $A_2$ . Duty cycle from 66.67% to 100% is the third area and is indicated as  $A_3$ . At least one switch must be ON because the input of the converter is current source, therefore  $A_1$  is forbidden. Figure 3 shows the working modes and equivalent circuits of each mode when the duty cycle is in  $A_2$  and are described for one switching period as under:

**Mode 1:**  $S_1$  start to conduct, direction of current at primary of  $T_1$  is altered and becomes negative;  $D_2$  starts to conduct and  $C_1$  begins to discharge.

**Mode2:** Gate pulse of  $S_3$  is removed and the primary current direction of  $T_3$  is altered and becomes positive value;  $D_5$  starts to conduct and  $C_6$  begins to discharge.

**Mode3:**  $S_2$  start to conduct, direction of current at primary of  $T_2$  is altered and becomes negative;  $D_4$  starts to conduct and  $C_3$  begins to discharge.

**Mode4:** Gate pulse of  $S_1$  is removed and the direction of current at primary of  $T_1$  is altered and becomes positive;  $D_1$  starts to conduct and  $C_2$  begins to discharge.

**Mode5:**  $S_3$  start to conduct, direction of current at primary of  $T_3$  is altered and becomes negative;  $D_6$  starts to conduct and  $C_5$  begins to discharge.

**Mode6:** Gate pulse of  $S_2$  is removed and direction of current at primary of  $T_2$  is altered and becomes positive;  $D_3$  starts to conduct and  $C_4$  begins to discharge.

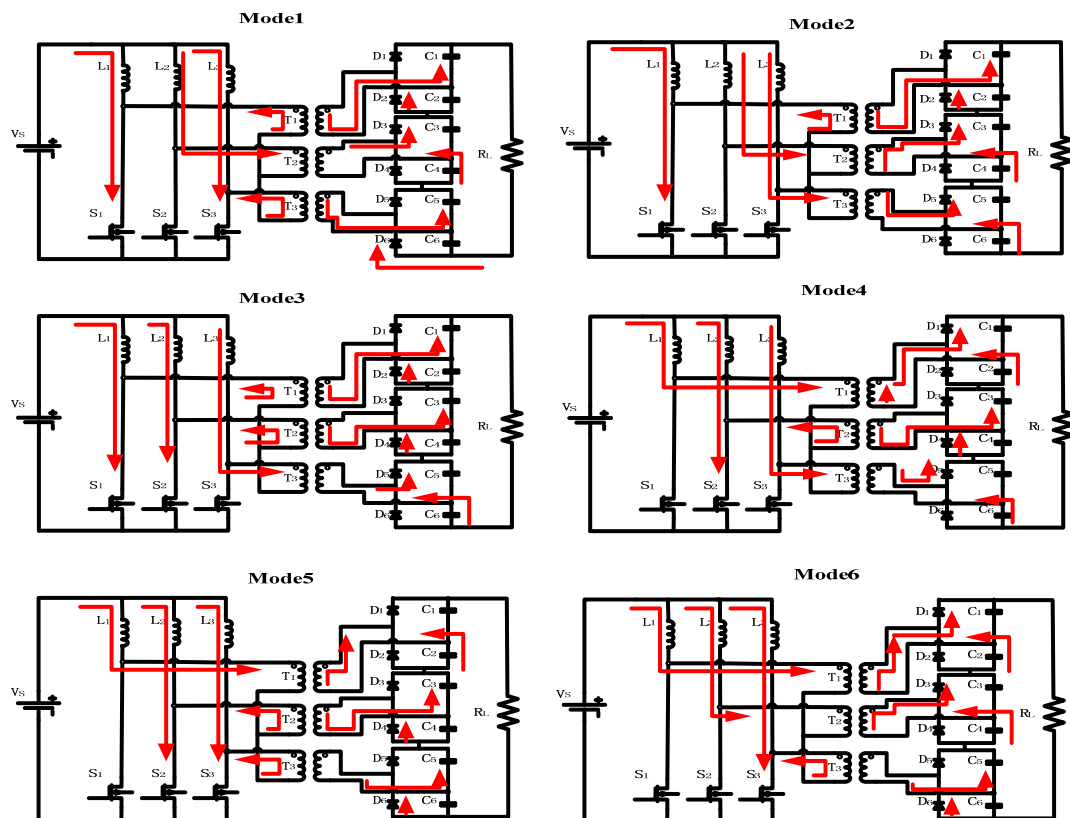


Figure 3. Shows the Working Modes and Equivalent Circuits When Duty Cycle Is In  $A_2$

**Mode 1:**  $S_1$  start to conduct, the primary current of  $T_1$ ,  $T_2$  and  $T_3$  become to zero; The diodes of the second sides do not conduct and the capacitances of the second sides begins to discharge together.

**Mode2:** Gate pulse of  $S_2$  is removed and direction of current at primary of  $T_2$  is altered and becomes positive; at the same time, the primary current direction of  $T_1$  and  $T_3$  is altered and becomes negative;  $D_2$ ,  $D_3$  and  $D_4$  starts to conduct.

**Mode3:** Gate pulse of  $S_3$  is removed direction of current at primary of  $T_3$  is altered and becomes positive;  $D_2$ ,  $D_3$  and  $D_5$  start to conduct.

**Mode4:**  $S_2$  start to conduct, direction of current at primary of  $T_2$  is altered and becomes negative;  $D_2$ ,  $D_4$  and  $D_5$  start to conduct.

**Mode 5:**  $S_3$  start to conduct, the primary current of  $T_1$ ,  $T_2$  and  $T_3$  become to zero; The diodes of the second sides do not conduct and the capacitances of the second sides begins to discharge together.

**Mode6:** Gate pulse of  $S_1$  is removed and the primary current direction of  $T_1$  is altered and becomes positive value; at the same time, the primary current direction of  $T_2$  and  $T_3$  is altered and becomes negative value;  $D_1$ ,  $D_4$  and  $D_6$  start to conduct.

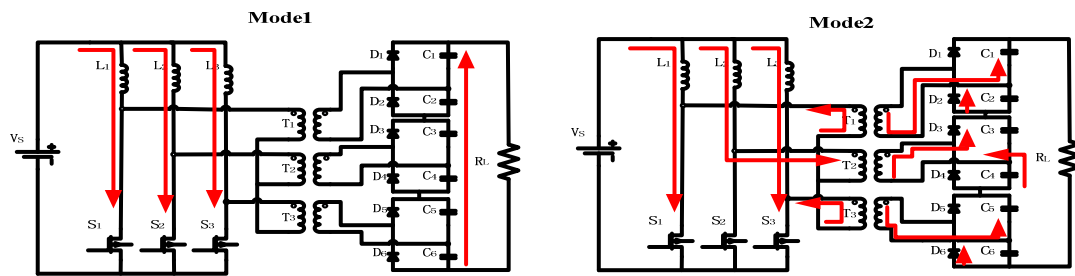
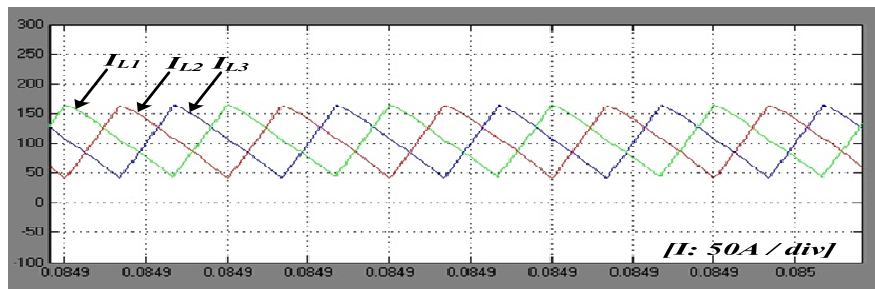


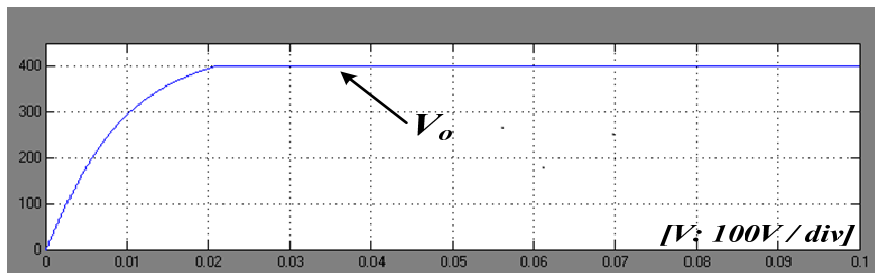
Figure 4. Shows the Working Modes and Equivalent Circuits When Duty Cycle Is In  $A_3$

### 3. Simulation and Experimental Results

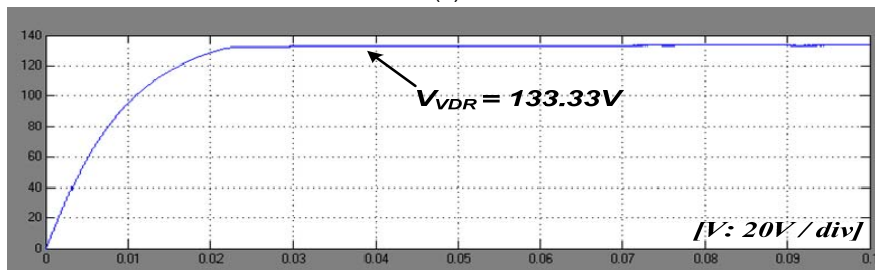
Multiphase step-up DC-DC converter topology along with the voltage doubler rectifiers has been simulated on SimPower tool of MATLAB in order to obtain high output voltage. The specifications for simulation model of above said configuration are shown in Table 1. The simulation results are represented in Figure 5, whereas Figure 5(a) shows input inductor currents, Figure 5(b) represents output voltage and Figure 5(c) shows voltage at each voltage doubler rectifier.



(a)



(b)



(c)

Figure 5. Simulation Results (a) Inductor Currents; (b) Output Voltage or load Voltage; (c) Voltage at each VDR

Table 1. Specifications for Simulation

Parameter	Value
$P_o$	5KW
$V_{in}$	50V
$V_o$	400V
F	40KHz
$L_1$ to $L_3$	1 $\mu$ H
$C_1$ to $C_6$	8 $\mu$ F

Table 2. Specifications for Experiment

Parameter	Value
$P_o$	2KW
$V_{in}$	35V
$V_o$	200V
F	40KHz
$L_1$ to $L_3$	160 $\mu$ H
$C_1$ to $C_6$	45 $\mu$ F

2KW Laboratory prototype has been design to verify the working mode and the control strategy of proposed DC-DC converter topology. The specifications for experimental model of above said conFigureuration are shown in Table 2.

Figure 6 shows the experimental results for the input inductors and output voltage.

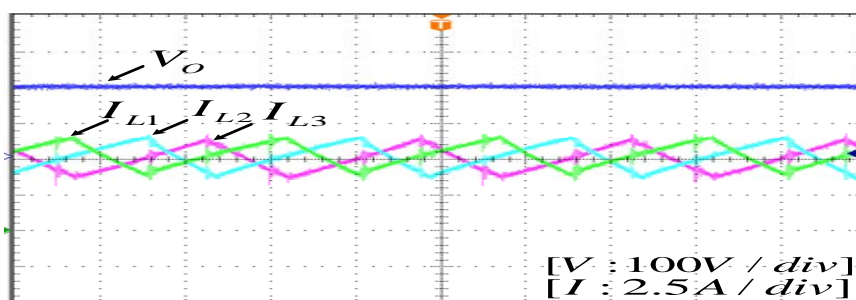


Figure 6. Experimental Results of Output Voltage and Inductor currents

#### 4. Conclusion

High output voltage based multiphase step-up DC-DC converter topology with voltage doubler rectifiers has been proposed in this paper. In the proposed topology two loop control strategy has been applied. Based on simulation and experimental results, the working modes of above said topology are explained in detail. Moreover, it can also be concluded that the high output voltage can be obtained by connecting series voltage doubler rectifiers at the secondary side of high frequency transformers. The converter topology is useful where the input DC voltage is less than the output DC voltage.

#### References

- [1] H Cha, J Choi, B Han. *A new three-phase interleaved isolated boost converter with active clamp for fuel cells*. Proc. IEEE PESC. 2008; 1271–1276.
- [2] SVG Oliveira, I Barbi. *A three-phase step-up DC-DC converter with a three-phase high frequency transformer*. Proc. IEEE ISIE. 2005; 571–576.
- [3] H Cha, J Choi, P Enjeti. *A three-phase current-fed DC/DC converter with active clamp for low-DC renewable energy sources*. *IEEE Transactions on Power Electronic*. 2008; 23(6): 2784–2793.
- [4] S Lee, S Choi. *A three-phase current-fed push-pull DC-DC converter with active clamp for fuel cell applications*. Proc. APEC. 2010; 1934–1941.
- [5] RL Andersen, I Barbi. *A three-phase current-fed push-pull DC-DC converter*. *IEEE Transactions on Power Electronic*. 2009; 24(2): 358–368.
- [6] AR Prasad, PD Ziogas, S Manias. *Analysis and design of a three phase offline DC-DC converter with high-frequency isolation*. *IEEE Transacation and Industrial Applications*. 1992; 28(4): 824–832.
- [7] D Souza Oliveira Jr, I Barbi. *A three-phase ZVS PWM DC/DC converter with asymmetrical duty cycle for high power applications*. *IEEE Transactions on Power Electronic*. 2005; 20(2): 370–377.
- [8] DS Oliveira Jr, I Barbi. *A three-phase ZVS PWM DC/DC converter with asymmetrical duty cycle associated with a three-phase version of the hybride rectifier*. *IEEE Transactions on Power Electronic*. 2005; 20(2): 354–360.
- [9] H Kim, C Yoon, S Choi. *A three-phase zero-voltage and zero current switching DC-DC converter for fuel cell applications*. *IEEE Transactions on Power Electronic*. 2010; 25(2): 391–398.

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- [10] J Lai. *A high-performance V6 converter for fuel cell power conditioning system*. Proc. IEEE VPPC 2005; 624-630.
- [11] K Fathy, H Lee, T Mishima, M Nakaoka. *Boost-half bridge single power stage PWM DC-DC converter*. Proc. IEEE PECon. 2006; 426-431.
- [12] Ahmad SS, Taufiq, Abd Jaafar S, Abdul Halim MY. *Simulation and Implementation of Interleaved Boost DC-DC Converter for Fuel Cell Application*. *International Journal of Power Electronics and Drive System (IJPEDS)*. 2011; 1(2): 168-174.
- [13] C Kim, G Moon, S Han. *Voltage doubler rectified boost-integrated half bridge (VDRBHB) converter for digital car audio amplifiers*. *IEEE Transactions on Power Electronic*. 2007; 22(6): 2321-2330.
- [14] H Watanabe, H Matsuo. *A novel high-efficient DC-DC converter with 1V/20A dc output*. Proc. IEEE INTELEC. 2002; 34-39.
- [15] J Zeng, J Ying, Q Zhang. *A novel DC-DC ZVS converter for battery input application*. Proc. IEEE APEC. 2002; 892-896.
- [16] Changwoo Yoon, Joongeun Kim, Sewan Choi. *Multiphase DC-DC Converters Using a Boost-Half-Bridge Cell for High-Voltage and High-Power Applications*. *IEEE transactions on Power electronic*. 2011; 26(2): 381-388.
- [17] D Choi, B Lee, S Choi, C Won, D Yoo. *A novel power conversion circuit for cost-effective battery-fuel cell hybrid systems*. *Journal Power Sour*. 2005; 152: 245-255.
- [18] VV Subrahmanya Kumar Bhajana. *Simulation Based Performance analysis of Active Clamp DHB ZVZCS Bidirectional DC-DC converter for Low Power Applications*. *International Journal of Power Electronics and Drive System (IJPEDS)*. 2012; 2(3): 345-352.