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Implementation of flow control over wirelessHART sensor network using wirelessHART adaptors

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Article Info

ABSTRACT

Despite the advantages of the industrial wireless standards such as WirelessHART, ISA100.11a and Wireless Networks for Industrial Automation-Process Automation (WIA-PA), their application still faces a lot of challenges especially when it comes to interfacing with the real plant. This is due to lack of adequate infrastructures such as interfacing circuitry to establish communication between the WirelessHART nodes and the actuators and sensors. Therefore, this paper presents the application of locally developed WirelessHART adaptors for flow process control. The adaptors serve as an interface between the WirelessHART network and the sensor and actuator of the plant. Experimental results of the controllers compared showed that wireless control is possible using the developed adaptors.

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1. INTRODUCTION

Recent advances in wireless sensor networks (WSNs) have attracted keen interest by researchers [1-5]. This is because in wireless communication technology, data and information are transmitted without the need of physical medium. Therefore, the technology brings about the benefits of reduce cabling as well as space saving among others. Despite these advantages, its deployment in the monitoring and control industry was sluggish. This was mainly due to lack of suitable industrial standards and interfaces such as wireless-Adaptor and interfacing software. Recent advances in the technology has led to the emergence of industrial wireless standards such as WirelessHART, WIA-PA and ISA100 wireless [6-9]. These standards are specifically designed for industrial monitoring and control applications in order to solve the problem of cumbersome cabling.

The WirelessHART, being based on the traditional HART protocol, has an edge over its counterpart with millions of HART-enabled devices already installed globally. However, applying the technology for control comes with the challenges of lack of adequate infrastructures such as WirelessHART adaptors. In an attempt to solve the problem of interfacing the WirelessHART network to the field elements such as sensors, transducers and actuators, Some prominent companies in the field, like EMERSON, AWIATECH, invested huge amount of money to create their own solution of wireless system between control plants, but these products are usually very expensive and are having proprietary issues. In a related development, a simple and inexpensive WirelessHART Adaptor was developed in [10] for process monitoring.

Therefore, in this paper, the developed adaptor is extended for implementation on a flow control. For this purpose, few controllers including those proposed in [11], [12] and [13] will be implemented for flow control on a Pilot Plant.

The rest of the paper is organized as follows: In Section 2, the brief on the complete WirelessHART control set-up including interfaces is given. Section 3 gives the brief of the compared controllers while Section 4 presents and discusses the results. Lastly, conclusion is provided in Section 5.

2. WIRELESSHART FLOW CONTROL LOOP SET-UP

This section will first describe the complete experimental set-up including the process and instrumentation (P&ID) diagram of the plant. Then, brief description of the adaptors/interface will be given at the second part.

2.1. Complete System Set up

The selected controllers will be implemented on a PcA SimExpert Flow Control and Calibration Process Mobile Pilot plant stationed at Block 23, Universiti Teknologi PETRONAS. The complete experimental set-up is shown in Figure 1 while the block diagram representation is given in Figure 2. Furthermore, the P&ID of the plant is shown in Figure 3. The plant consist of a buffer tank and a calibration tank that are connected in series. The objective of the plant is to transfer fluid from the buffer tank to the calibration tank at a controlled flow rate. To achieve this, two pumps and a pneumatic valve are used. In this experiment, the measured value is water level while the manipulated value is the flow rate in m3/s. As seen in the figure, PIC110C is the main feedback controller while FT110C and CV110 are the flow transmitter and the control valve respectively. P101 and P201 are the pumps for buffer tank VE100 and calibration tank VE200 respectively. In the same vein, LS101 and LS201 are the respective level sensors connected to P101 and P201 for the control of overflow.

To achieve wireless control of the flow rate by controlling the opening of CV110 and to obtain flow measurement, the controller is implemented in Simulink environment in the host computer interfaced with Python to export the control action into the gateway. The control signal or manipulated variable (MV) is now received by the valve through a developed WirelessHART Adaptor. Measurements based on the process variable (PV) are received from the sensor via WirelessHART adaptor into the gateway and then the controller.
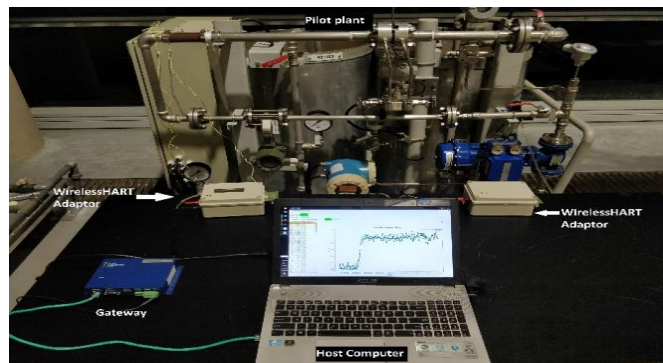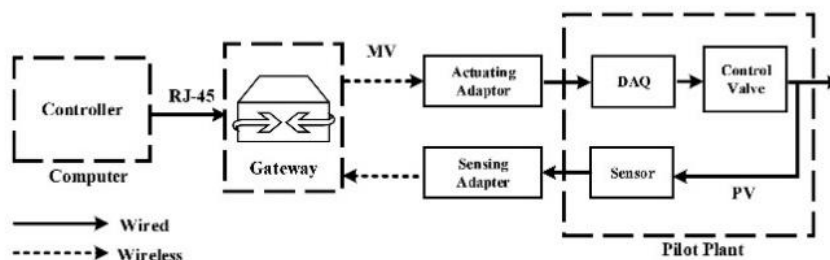

Figure 1. Complete experimental set-up

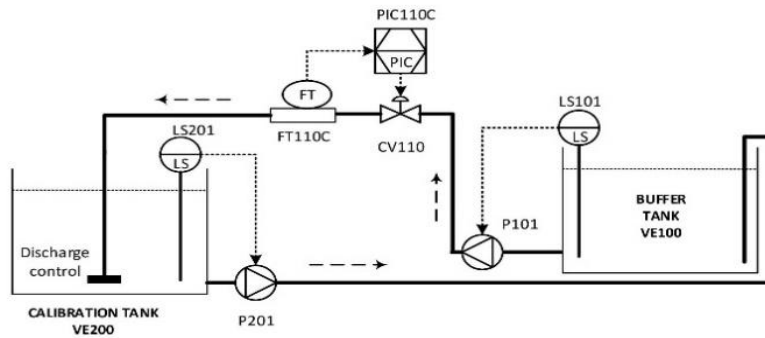Figure 2. Block diagram representation of the experimental set-up

Figure 3. Simplified P&ID diagram

2.2. Adaptors / System Interface
2.2.1 WirelessHART Sensing Adaptor

As shown in the block diagram of Figure 4, the sensing module is primarily formed by Arduino MEGA 2560 micro-controller board and a WirelessHART mote (DC9003A-C) from Linear Technology. In addition, an LCD display is added to the design in order to display real time sensor reading. This can be seen in the circuit implementation of the adaptor given in Figure 5. The pin-out connections between the micro-controller and the WirelessHART mote is given in Table 1.
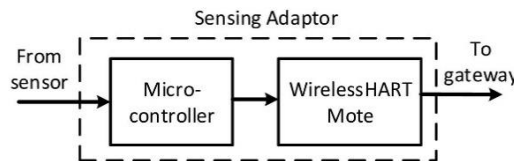

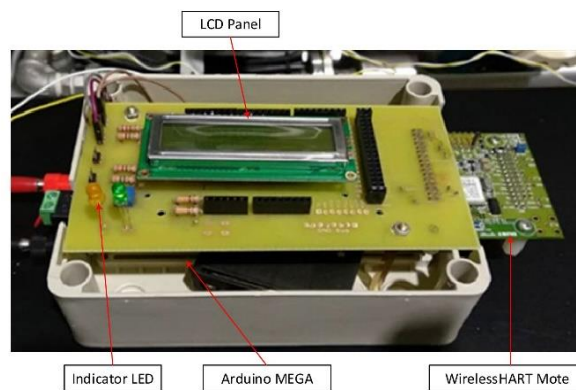Figure 4. Block diagrams of Sensing WirelessHART adaptor

Figure 5. Sensing WirelessHART adaptor

Table 1. Mote/Micro-controller pin-out

DC9003A-C	Mega 2560
GND	GND
VBAT	3.3V
RX	TX1
TX CTSn	GND
RX CTSn	3.3V

2.2.2 WirelessHART Actuation Adaptor

Similar to the sensing module, the actuating module is formed by Arduino MEGA 2560 micro-controller board and a WirelessHART mote (DC9003A-C). In addition, a 4-20mA T click board is used to convert the signal from the micro-controller into a 4-20mA signal. The main function of this module is to channel the control action generated by the host application to the valve in order to adjust the degree of actuation. Block diagrams as well the circuit implementation of the adaptor are shown in Figure 6 and 7 respectively. Furthermore, the pin-out connections between the micro-controller and the WirelessHART mote is similar to that of sensing adaptor and is given in Table 2.
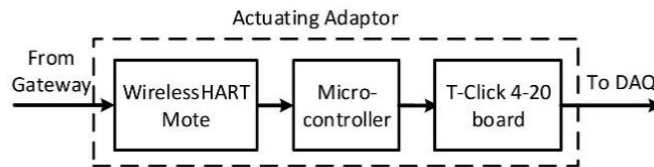

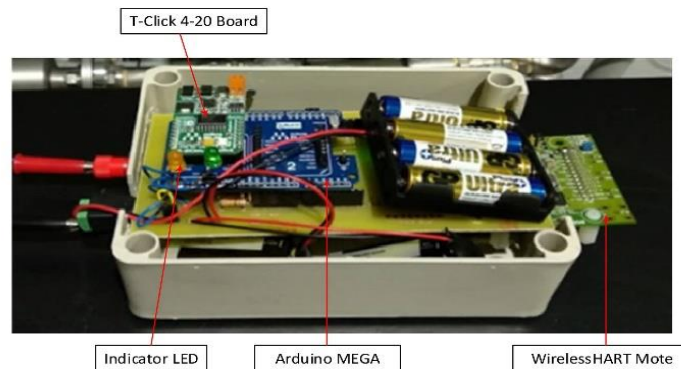Figure 6. Block diagrams of Actuating WirelessHART adaptor

Figure 7. Actuating WirelessHART adaptor

The control valve in the plant is the Spirax Sarco pneumatic valve. Thus, the plant is equipped with a current to pressure converter (I/P Converter) that converts the 4-20mA current into equivalent pneumatic pressure value. The details of the I/P converter are shown in Table 2.

Table 2. I/P Converter Specifications

Spirax Sarco EP5	Electro pneumatic Positioner
Input Signal	4-20mA
Output Signal Range	0-100% supply air pressure
Supply air pressure	1.4-6.0 bar
Voltage Rating	5V (Min) 24V (Max)

3.   CONTROL STRATEGIES

This section will briefly discuss the controllers to be compared. The controllers compared are PI, Smith predictor, Fuzzy PID, Setpoint Weighting and Filtered Predictive PI. Some of these controllers are reported in our earlier works in [11, 13]. For easy understanding of the controllers, consider the single loop WirelessHART network shown in Fig. 8. To facilitate analysis, the delays can be lumped as total network-induced delay τ_N as given in (1).

$$\tau_N = \tau_{CA} + \tau_{SC} \tag{1}$$

where τ_{CA} and τ_{SC} are the controller-to-actuator and sensor-to-controller delays respectively, $G(s) = G_n(s)e^{-sL_P}$ is the plant model. Assuming there is commutativity, the process deadtime (L_P) as given in the figure can be added to the network delay which now gives the total closed loop delay as in (2). These equations will be required for the design of the controllers.
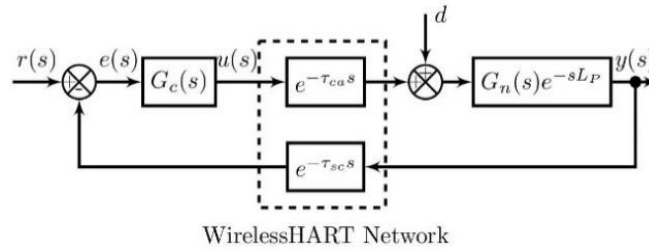
$$L = \tau_N + L_P \tag{2}$$

Figure 8. Network delay representation in a single loop WirelessHART networked control system

3.1. PI, Smith Predictor and Fuzzy PID

The controller structure used for both PI and Smith predictor [14] is given in (3).

$$G_c(s) = K_c \left(1 + \frac{1}{T_i s}\right) \tag{3}$$

where K_c and T_i are the controller gain and time constants respectively. On the other-hand, the Fuzzy PID control structure adopted in this work is shown in Figure 9.
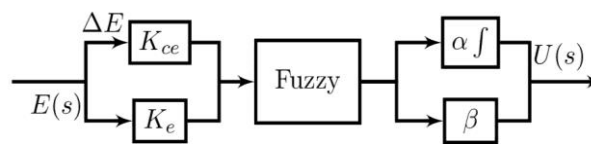

Figure 9. Fuzzy PID Structure

As seen from the structure, the controller has two inputs (error and change in error) and one output. The input scaling factors (SFs) K_e and K_{ce} are the respective error and error change gains, while the output SFs α and β are the control gains. The output U(s) of the controller is given in (4):

$$u(t) = K_p + K_i \int e(t)dt + K_d \frac{de(t)}{dt} \tag{4}$$

In (4), Kp, Ki, and Kd are the proportional, integral, and derivative constants respectively. They are related to the fuzzy PID gains as $K_p = \alpha K_{ce} + \beta K_e$, $K_i = \alpha K_e$, $K_d = \beta K_{ce}$. Where, Ke is the input error scaling factor, Kce is the input error change scaling factor, α and β are the output scaling factors

3.2. Filtered Predictive PI (FPPI)

The FPPI structure is given in shown in Figure 10. The difference between the conventional PPI structure and the FPPI is the inclusion of the filter term F(s) which will help curtail the effect of noise and oscillation induced by higher order systems and stochastic nature of the network. It should be noted that the design of PPI is based on FOPDT systems [15-16].

$$U(s) = \left(K_c E(s) + \frac{1}{1+sT_i} e^{-sL} U(s)\right) F(s) \tag{5}$$
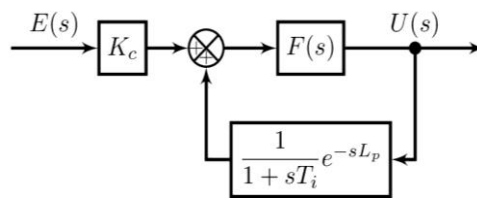
where F(s) is a filter transfer function.

Figure 10. Implementation of FPPI controller

3.3. Setpoint Weighting (SW)

Consider the reference signal r(s) of Figure 8, using setpoint weighting function fr(s) of (6), this reference signal is varied from r(s) to $\tilde{r}(s)$. The implementation of (6) is shown in Figure 11. This allows for the 2-DoF ability of both good setpoint tracking and disturbance rejection of the controller. The complete design procedure for this approach can be found in our earlier work reported in [13, 17-21].

$$f_r(s) = \frac{\tilde{r}(s)}{r(s)} = G_r(s) + \tilde{G}_{yr}(s)\left(e^{\tilde{r}s} - G_r(s)\right) \tag{6}$$

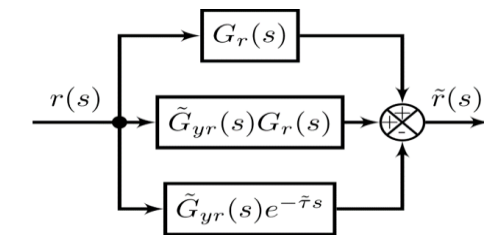where Gr(s) is the setpoint regulating feed-forward controller.

Figure 11. Structure of the general SW function fr(s)

4. RESULTS AND ANALYSIS

This section is divided into two parts. In the first part, various controller parameters and flow process plant model will be presented while the second part of the section will present the result comparison.

4.1. Plant model and controllers Parameters

The controllers were tested on a pilot flow control system of Figure 1. The model of the system is given in the transfer function of (7). The model was obtained using empirical modelling of the plant similar to that reported in [22] and [23]. The various controller parameters for the plant are given in Table 3. The Fuzzy PID and FPPI parameters were obtained through tuning with optimization algorithm on the model of the plant. Furthermore, the rule base table for the fuzzy PID based on [24] and [25] is given in Table 4. The definitions of the entries to the table are given as: Z for "Zero" while NS, NM, and NB stands for "Negative small", "Negative medium" and "Negative big" respectively. Others are PS, PM and PB which stands for "Positive small", "Positive medium" and "Positive big" respectively.

$$G = \frac{0.58}{0.26s+1} e^{-0.25s} \tag{7}$$

4.2. Result Comparison

In this experiment, the data sampling time in MATLAB is set as 0.5 seconds while the update rate between the gateway and field devices (motes/adaptors) is set as the 4s which is at the middle of 1s for fast

piping and 8s for optimal battery performance. The disadvantage of the fast piping is that it drains the batteries of field devices faster. The duration of the experiment for each controller tested is 500 seconds. This is due to the capacity of the tanks VE100 and VE200. The target was set at 1.5m3/hour. It should be noted that the valve position at 100% delivers a maximum of 2.5m3/hour.

Table 3. Various controller parameters

Controller	Parameters
PI & Smith	4-20mA
FPPI	0-100% supply air pressure
SW	1.4-6.0 bar
Fuzzy PID	5V (Min) 24V (Max)

Table 4. Fuzzy PID rule base

E\ΔE	NB	NM	NS	Z	PS	PM	PB
PB	Z	PS	PM	PB	PB	PB	PB
PM	NS	Z	PS	PM	PB	PB	PB
PS	NM	NS	Z	PS	PM	PB	PB
Z	NB	NM	NS	Z	PS	PM	PB
NS	NB	NB	NM	NS	Z	PS	PM
NM	NB	NB	NB	NM	NS	Z	PS
NB	NB	NB	NB	NB	NM	NS	Z

Result comparison of the PI, FPPI, SW, Fuzzy PID and Smith predictor controllers is shown in Fig. 12. The numerical information of the result as obtained from the figure is given in Table 5. From the figure, it can be clearly seen that FPPI controller has faster response with a rise time of around 34s. This is followed by PI and SW with 46.77s and 56.22s each. The slowest are the fuzzy and Smith predictor with respective rise times of 72.27 and 164.73s. Although the smith predictor recorded the least overshoot of 6.7%, it is however very slow in response and experiences glitch at around 150s during the experiment. Comparing overshoots of the controllers against the PI, it can be seen that there is a significant difference between the 31.33% of the PI and the 6.7, 11.33, 14.67 and 16% of the Smith Pred., SW, Fuzzy and FPPI respectively. In a nutshell, all the other controllers have outperformed the PI in terms of overshoot while the FPPI has an added advantage of rise time over the PI.
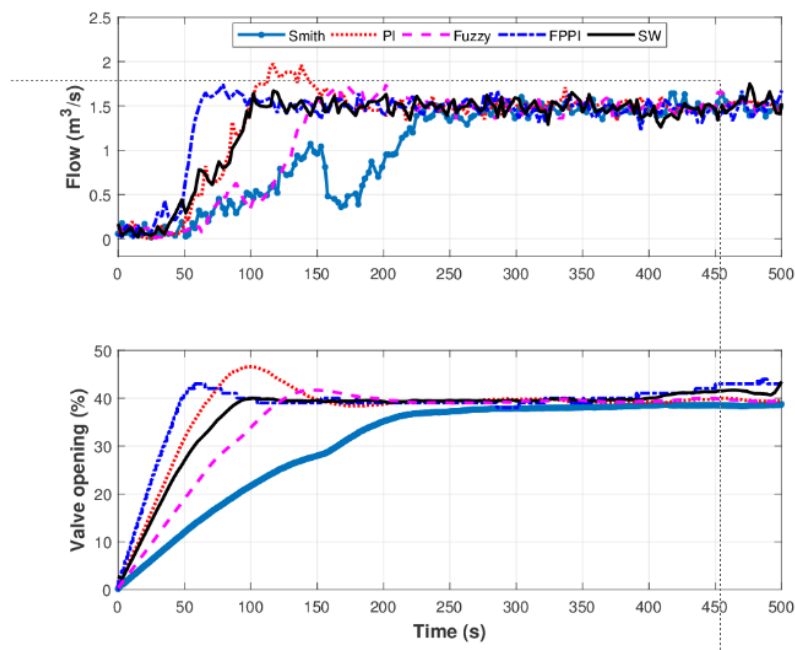

Figure 12. Comparison of various developed controllers for flow control

Table 5. Performance of controllers for pilot plant

Controller	System's Performance	
	Rise Time (s)	Overshoot (%)
SW	56.22	11.33
FPPI	33.95	16.00
Fuzzy PID	72.27	14.67
PI	46.77	31.33
Smith Pred.	164.73	6.70

The respective characteristics of each controller as seen from its response is also manifested through the control signals. While the signal of FPPI rose steadily to 40% in less than 50s, the Smith predictor is the most sluggish by reaching the 40% mark at around 200s. It should also be noted that, the signal of SW despite being slower than both FPPI and PI, does not go beyond the 40% position of the valve as compared to the FPPI, PI and Fuzzy PID. This is responsible for its lower overshoot compared to the Fuzzy PID, FPPI and PI. Thus, after comparing all the results, it can be succinctly stated that the SW, FPPI and fuzzy PID controllers have improved on both the PI and Smith predictor controllers.

5.   CONCLUSION

In this paper, the use of WirelessHART adaptors for control has been demonstrated. The sensing and actuating adaptors employed here were developed locally using inexpensive components. Thus, these adaptors permitted the actualization of wireless control. Furthermore, experimental results of the controllers compared shows that despite the network induced delay, improved performance of the PI and Smith predictor approaches can be achieved through the use of FPPI, SW and fuzzy PID strategies.

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