

Advanced optimal by PSO-PI for DC motor

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ABSTRACT

Interest in the drive system, it used in many applications is increasing day by day. Therefore, many researchers have focused on the analyses, design and control of these systems. In this study, Optimal for Dc motor in drive system control strategy proposed for PSO-PI and fuzzy logic controller (FLC) based Dc motor in drive system. In order to test dynamic performance of PSO-PI based drive System, simulation study carried out by MATLAB/Simulink. The results obtained from the PSO-PI based drive system are not only superior in the rise time, settling time and overshoot but can prevent from voltage and has improved power quality.

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1. INTRODUCTION

Increasing orientation for the use of Dc motor in industry for many electrical applications, EV, Elevator, Robot .etc. [1-3]. A way controller (PI) in addition to the controller integral relative formulated and implemented, using speed control drive system and a pilot phase. While the new strategy promotes traditional PI control performance to a large extent, and proves to be a model-free approach completely, it also keeps the structure and features of a simple PI controller [4-5]. The use of fuzzy logic (FL) instead of Fuzzy-PI control to improve the performance of engines. To control the speed of motor using fuzzy logic (FL) approach leads to a speed control to improve the dynamic behavior of the motor drive system and immune disorders to load and parameter variations [6-7]. In the drive systems, and gains from the traditional can't usually be set in proportion-integral (PI) controller because of mechanical resonance. As a result, performance degradation control. In our work described in this paper, have been adopted and fuzzy logic controller (FLC) for use in drive systems in order to improve the performance of the DC Motor control. The proposed FLC has been compared with traditional PI control with respect to the DC Motor of response [8-9]. Simulation and experimental results have proved that FLC was proposed is superior to the traditional PI. This FLC can be a good solution for the high-performance DC Motor systems. A modern approach to control the speed of load using particle swarm optimization (PSO) to improve the algorithm parameters observer PI-. Simulate the system under different operating year conditions is prepared and the experimental setup. Use PSO algorithm and optimization make a powerful engine, with faster response and higher resolution dynamic and sensitive to load variation [10].

2. DC MOTOR

The Motor, It is a device converts electrical energy to mechanical energy or a device that produces rotational force which have been developed for different specific purposes that have brought about one of advance in the fields of technology. It is running by DC source [11-12]. The Motor Classification to, AC (alternating current) electrical motors, DC (Direct Current) motor and Special Types of Motor. DC Motor, the electrical port include the supply voltage E and current I is given to the input port and the mechanical port include torque T and speed ω from output port [13-15]. Model of DC motor as shown in Figure 1.

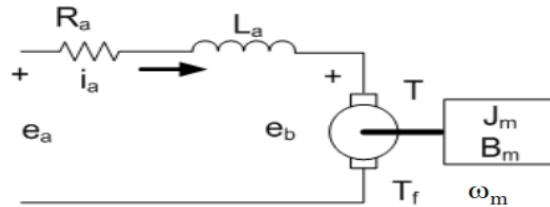


Figure 1. Model of DC Motor

2.1. Mathematical Model of DC Motor

In Figure 1, the model of DC motor include the Electrical part and Mechanical part which have Electromechanical Relationships to analysis mathematical model of DC motor by the following (1-11) [16-17]:

$$T = K_T \cdot i_a - T_f N - m \quad (1)$$

$$e_a = i_a \cdot R_a - e_b V \quad (2)$$

$$e_b = K_e \cdot \omega_m V \quad (3)$$

$$P = \omega_m \cdot T V \quad (4)$$

$$\omega_m = \frac{K_T \cdot e_a - (T - T_f) \cdot R_a}{K_T \cdot K_e} \quad (5)$$

$$\omega_m = \frac{e_a - i_a \cdot R_a}{K_e} \quad (6)$$

$$\omega_m = \frac{K_T \cdot e_a + (T_f) \cdot R_a}{K_T \cdot K_e} \quad (7)$$

Electrical Equations and Mechanical Equations (Electromechanical Relationships)

Electrical:

$$e_a(t) = R_a \cdot i_a(t) + L \frac{di_a(t)}{dt} + e_b(t) \quad (8)$$

Mechanical:

$$T(t) = J_m \frac{d\omega_m(t)}{dt} + B_m \cdot \omega_m(t) \quad (9)$$

Electromechanical:

$$e_b(t) = K_E \cdot \omega_m(t) \quad (10)$$

$$T(t) = K_T \cdot i_a(t) \quad (11)$$

Laplace Transform of:

Electrical;

$$E_a(s) = L \cdot s \cdot I_a(s) + R_a \cdot I_a(s) \tag{12}$$

$$I_a(s) = \frac{1}{L \cdot s \cdot R_a} [E_a(s) - E_b(s)] \tag{13}$$

Mechanical;

$$T(s) = (j_m \cdot s + B_m) \cdot \omega_m(s) \tag{14}$$

$$\omega_m(s) = [1 / (j_m \cdot s + B_m)] \cdot T(s) \tag{15}$$

Electromechanical;

$$e_b(s) = K_E \cdot \omega_m(s) \tag{16}$$

$$T(s) = K_T \cdot i_a(s) \tag{17}$$

Where: $i_a(t)$ = armature current, $e_b(t)$ = back emf, $e_a(t)$ = armature terminal voltage, ω_m = motor speed (rad/sec), T = motor torque, T_f = static friction torque, R_a = armature resistance, L_a = armature inductance, J_m = rotational inertia, B_m = viscous friction, K_T = torque constant, K_e =back emf constant and P = shaft power.

2.2. Transfer Function of DC Motor

A transfer function for a dc motor that relates input voltage to shaft position. The transfer function for an armature controlled dc motor. Represent a mechanical load using a mathematical model. How negative feedback affects dc motor performance. In Figures 2 and 3, Block diagram of DC Motor with load and Block diagram of DC Motor with no load ($T_f=0$). A transfer function of DC motor by the following (18-24) [18-20]:

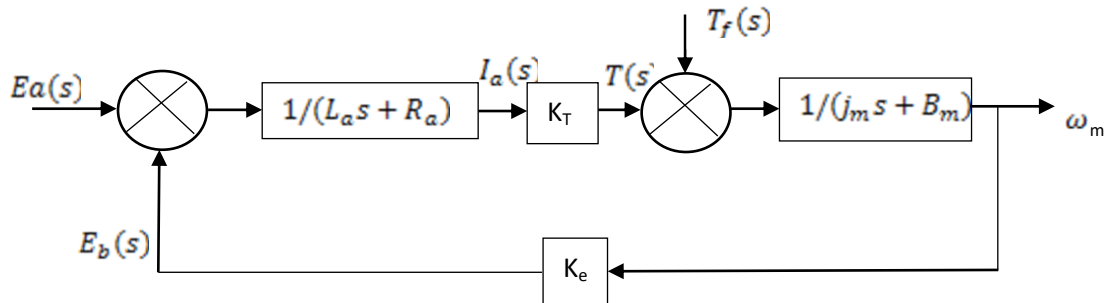


Figure 2. Block diagram of DC Motor with load

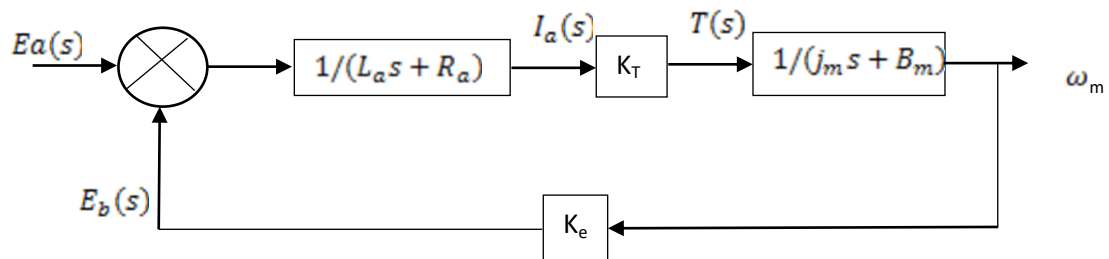


Figure 3. Block diagram of DC Motor with no load ($T_f=0$)

In (18-21), F.B, F.W and T.F formula:

$$\omega_m(s) = G(s) \quad (18)$$

$$G(s) = \frac{K_T}{(L_a \cdot s + R_a)(j_m \cdot s + B_m)} \quad (19)$$

$$\text{F.W formula: } E_a(s) = 1 + G(s) \cdot H(s) \quad (20)$$

$$\text{F.B formula: } H(s) = K_E \quad (21)$$

T.F formula:

$$\frac{\omega_m(s)}{E_a(s)} = \frac{G(s)}{1 + G(s) \cdot H(s)} \quad (22)$$

Simplify T.F formula:

$$\frac{\omega_m(s)}{E_a(s)} = \frac{\frac{K_T}{(L_a \cdot s + R_a)(j_m \cdot s + B_m)}}{1 + \left[\frac{K_T}{(L_a \cdot s + R_a)(j_m \cdot s + B_m)} \right] \cdot K_E} = \frac{K_T}{(L_a \cdot s + R_a)(j_m \cdot s + B_m) + K_T \cdot K_E}$$

Final T.F formula:

$$\frac{\omega_m(s)}{E_a(s)} = \frac{G(s)}{1 + G(s) \cdot H(s)} = \frac{K_T}{L_a j_m \cdot s^2 + (R_a j_m + B_m L_a) \cdot s + (K_T \cdot K_E + R_a \cdot B_m)} \quad (23)$$

Parameters of DC motor:

$\omega_m = 500 \text{ rad/s}$, $I_a = 2 \text{ A}$, $K_e = 0.06 \text{ V} \frac{\text{s}}{\text{rad}}$, $K_T = 0.06 \text{ N} \frac{\text{m}}{\text{A}}$, $T_f = 0.012 \text{ N} \cdot \text{m}$, $R_a = 1.2 \text{ ohm}$
 $j_m = 6.2 \cdot 10^{-4} \text{ N} \cdot \text{m} \cdot \text{s} / \text{rad}$, $B_m = 1 \cdot 10^{-4} \text{ N} \cdot \text{m} \cdot \text{s} / \text{rad}$, $L_a = 0.02 \text{ H}$.

$$T.F = \frac{\omega_m(s)}{E_a(s)} = \frac{16.13}{1 + 0.201s + 0.00333s^2} \quad (24)$$

2.3. Parameters Mathematical Results of DC Motor

Parameters:

$T_f = 0.012 \text{ N} \cdot \text{m}$, $R_a = 1.2 \text{ ohms}$, $K_e = 0.06 \text{ N} \cdot \text{m} / \text{A}$, $K_T = 0.06 \text{ V} \cdot \text{s} / \text{rad}$, $i_a = 2 \text{ A}$ and $\omega = 500 \text{ rad/s}$

Mathematical results:

By using: (1): $T = (0.06 \cdot 2) - 0.012 = 0.108 \text{ N} \cdot \text{m}$

(2): $e_a = (1.2 \cdot 2) + (0.06 \cdot 500) = 32.4 \text{ Volt}$

(4): $P = 500 \cdot 0.108 = 54 \text{ watt}$

(7): $\omega_m = (0.06 \cdot 32.4) + [(1.2 \cdot 0.012) / (0.06 \cdot 0.06)] = 536 \text{ rad/s}$

3. OPTIMIZATION AND CONTROLLER

Optimization and Controller (Classical controller type PI Controller, Expert System type FLC and Optimization type and PSO), Electric system driven is formulated by the motor.

3.1. PI Controller Modeling

In the PI controller DC Motor, the actual value is compared with the reference value and the error is the nth sampling interval as [20-21].

$$\omega_e[n] = \omega_r^*[n] - \omega_r[n] \quad (25)$$

The output of the DC Motor controller gives the reference. Hence the output of DC Motor controller at the nth sampling interval is:

$$T[n] = T[n-1] + K_p(\omega_e[n] - \omega_e[n-1]) + K_i \cdot \omega_e[n] \quad (26)$$

For constant air gap flux operation reference quadrature axis current is given as:

$$i_q^* = T[n]/K_t \tag{27}$$

The limiter is used to limit the maximum value of output of speed controller. The maximum machine rated current and device current of the converter dictate the limit. Block diagram of the PI controller as shown in Figure 4.

Where,

- $\omega_e[n]$ is speed error at nth instant,
- $\omega_r^*[n]$ is the reference speed at nth instant
- $\omega_r[n]$ is the actual machine speed at nth instant,
- $\omega_e[n-1]$ is the speed error at (n-1)th instant
- $T[n]$ is the reference torque at nth instant,
- $T[n-1]$ is the reference torque at (n-1)th instant
- K_p is proportional gain of the speed controller
- K_i is integral gain of the speed controller is reference quadrature axis current
- K_t is torque constant

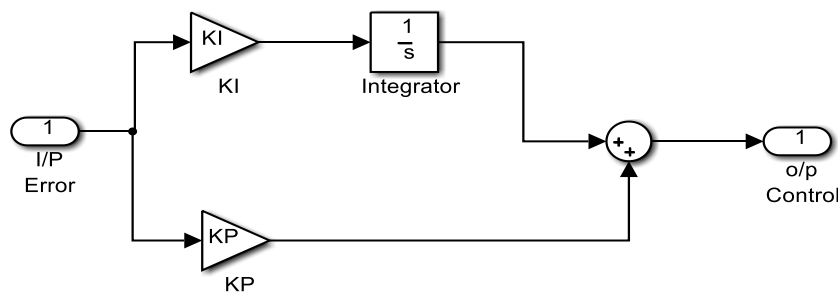


Figure 4. Block diagram of the PI controller

3.2. Fuzzy Logic Controller

Fuzzy logic controllers have the following advantages over the conventional controllers that they are cheaper to develop, they cover a wide range of operating conditions, and they are more readily customizable in natural language terms. In Mamdani type FIS the crisp result is obtained by defuzzification, in the Mamdani FIS can be used for both multiple input and single output and multiple inputs multiple outputs system as shown in Figure 5.

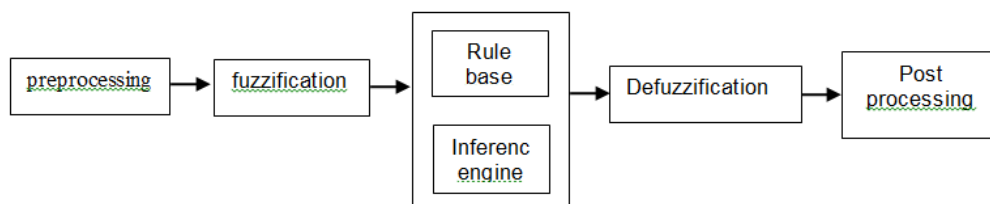


Figure 5. Arrangement of fuzzy logic controller

The usefulness of fuzzy logic controller is adopted especially in a complex and nonlinear system. The rules of conventional FLC are produced depend on the operator's experience or general knowledge of the system in a heuristic way. The thresholds of the fuzzy linguistic variables are usually chosen arbitrarily in the design process. An improper controller value leads to an adverse consequence, unstable mode, collapse and separation. This work propose PSO to design an Optimal Fuzzy Logic Controller OFLC, the optimized criteria is how to minimizing the transient state. The motor in a nonlinear system, time-change and complex system. The PI control is difficult to realize, which needs the accurate mathematical model with synthesizes the fuzzy control and PI control, the parameters can be adjusted deceptively. The FLC has better performance and robustness than conventional PI controller in the servo motor system. Block diagram of FLC- PI controller as shown in Figure 6. It can be overcome by the fuzzy logic controllers. PI control has good

robustness, simple in structure, and high reliability which suitable for a deterministic system which has accurate mathematical model. The control system performance rely on PI controller parameters [22-23].

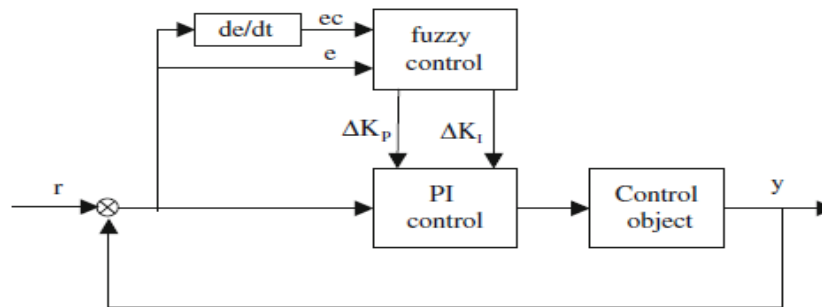


Figure 6. Block diagram of fuzzy PI controller

3.3. Particle Swarm Optimization

The biggest characteristic of PSO is in its simple structure, fast convergence, and its ability to prevent falling into a local optimum solution. At the same time, PSO is a random algorithm with a parallel structure. A uniform distribution is used to randomly create a particle swarm. Each particle represents a feasible solution to the problem, the particle swarm refers to the individual's best experience, and the group's best experience, and logically chooses the method it will move itself. After continuous iterations, the particle swarm will gravitate towards the optimum solution. For the i th particle and n -dimensional space can be represented as an (30), the best previous position of its particle is recorded as (31) [24-25]:

$$x_i = (x_{i,1}, x_{i,2}, \dots, x_{i,n}) \quad (28)$$

$$p_{best_i} = (p_{best_{i-1}}, p_{best_{i-2}}, \dots, p_{best_{i+n}}) \quad (29)$$

The velocity is an essential part of how PSO work so as modified velocity and position of each particle can be calculated using the current velocity and distance from ($p_{best_{i,d}}$) to (g_{best_d}):

$$V^{(It+1)}_{i,m} = W * V^{(It)}_{i,m} + c_1 * r * (p_{best_{i,m}} - x^{(It)}_{i,m}) + c_2 * r * (g_{best_{i,m}} - x^{(It)}_{i,m}) \quad (30)$$

$$x^{(It+1)}_{i,m} = x^{(It)}_{i,m} + V^{(It)}_{i,m} \quad (31)$$

Where n is the number of particles in a group; m is number of Dimension $m=1, 2, \dots$; It is a pointer of iterations (generations); W is an inertia weight factor; c_1, c_2 are acceleration constants were often set to be 1.2 according to past experiences; r is random value in the range between $[0,1]$; $V^{(It)}_{i,m}$: Velocity of particle no. i at iteration It , $x^{(It)}_{i,m}$: Current position of particle i at iteration It . G_{best_m} : Global best particle among all the particles in the population. In the above procedures, the parameter determined the resolution, or fitness, with which regions are to be searched between the present position and the target position, the inertia weight is set according to the (32).

$$W = w_{max} - \frac{w_{max} - w_{min}}{Iter_{max}} * Iter \quad (32)$$

Where: $Iter_{max}$ is the maximum number of iterations and $Iter$ is the current number of iterations.

4. ADVANCED IMPLEMENTATION FOR DC MOTOR

To use different control systems, like Classical PI Controller, Expert System Fuzzy Logic Controller and Optimization PSO Controller. It used to control for Implementation for DC Motor. The simulation model for DC motor as shown in Figure 7 and The simulation model as shown in Figures 8-11, by used all types to get the result and analysis it with compared to see the advanced implementation for DC Motor:

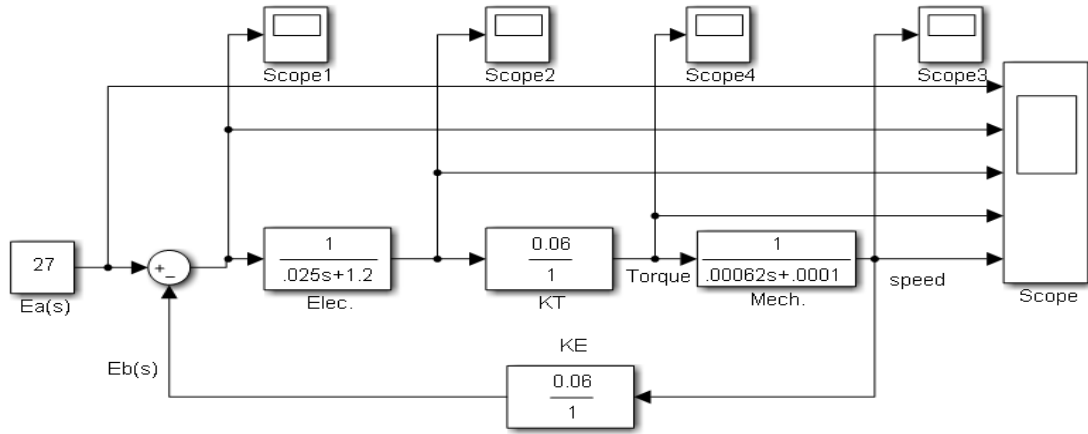


Figure 7. The simulation model for DC Motor

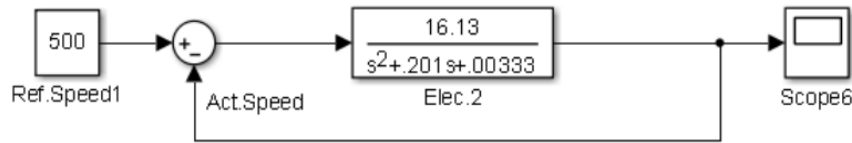


Figure 8. The simulation model for DC Motor without control

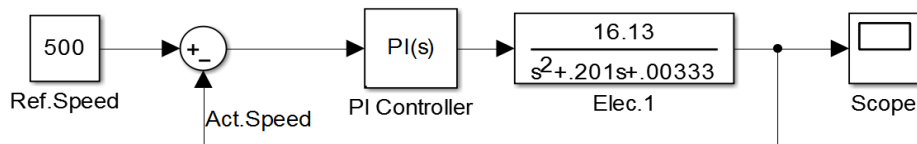


Figure 9. The simulation model for DC Motor with PI control

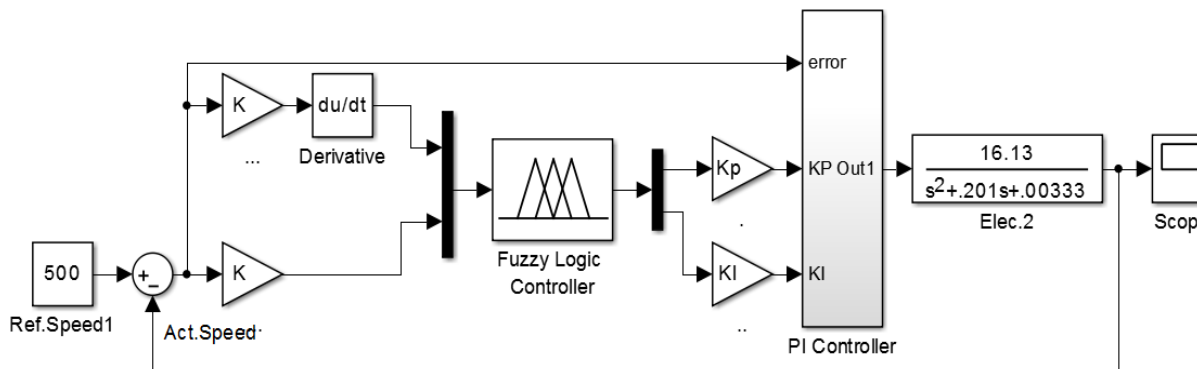


Figure 10. The simulation model For DC Motor with FLC

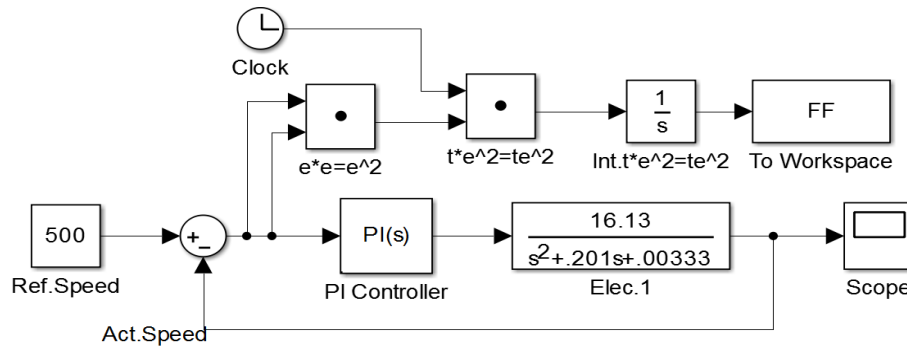


Figure 11. The simulation model For DC Motor with PSO-PI control

5. RESULTS AND ANALYSIS

5.1. The simulation Results

5.1.1 The Simulation Results for DC Motor with in and Without Controller

By used the simulation model for DC Motor without control in Figure 12 and with PI control to get the simulation results for DC Motor with controller in Figure 13.

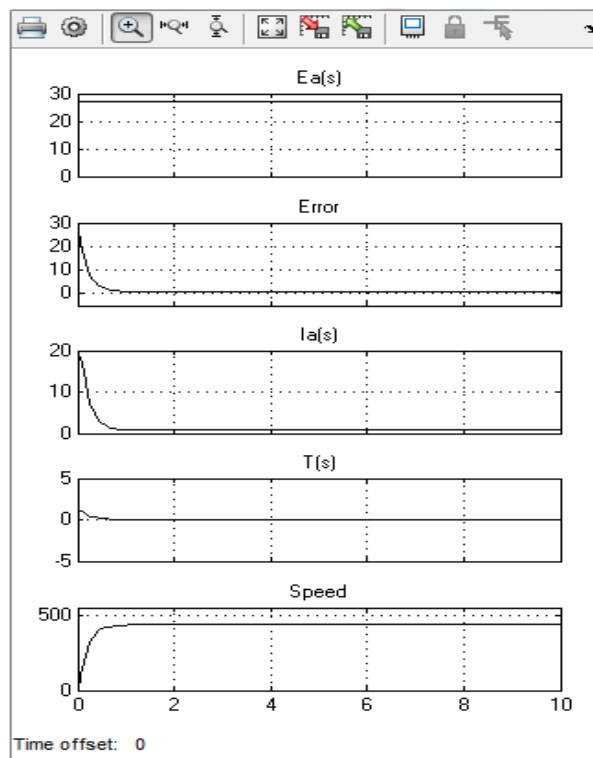


Figure 12. Simulation speed response for T.F of DC Motor without Controller

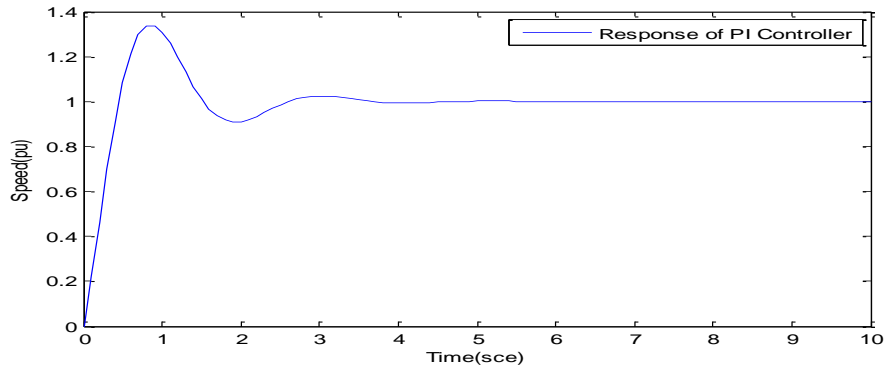


Figure 13. Simulation speed response for T.F of DC Motor with PI Controller

5.1.2 The simulation Results for Dc Motor with Optimization

By used the simulation model for Dc Motor with Optimization to get the simulation results for DC Motor with Optimization in Figures 14 and 15.

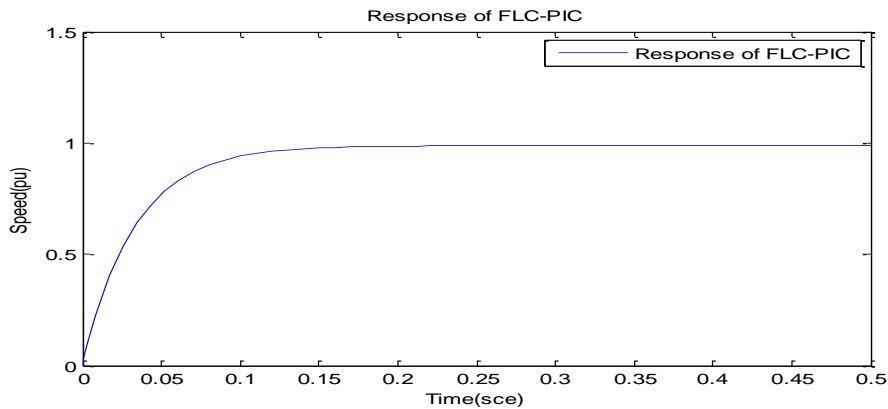


Figure 14. Simulation speed response for T.F of DC Motor with FLC_ PI Controller

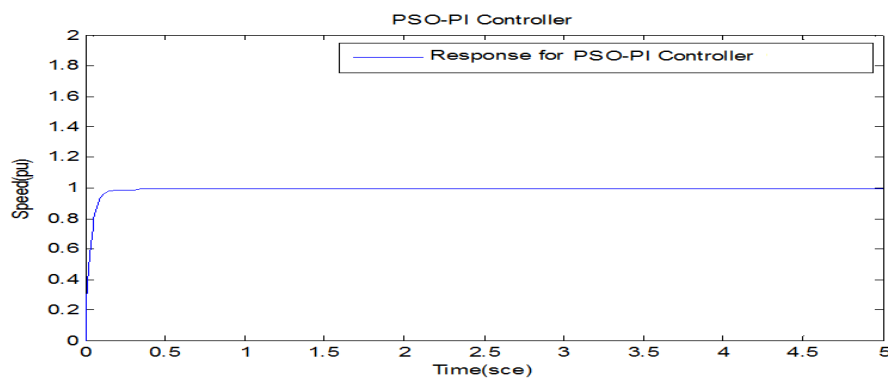


Figure 15. Simulation speed response for T.F of DC Motor with PSO_ PI Controller

5.2. Simulation Analysis

Final step, use different control systems, Like Classical PI Controller, Expert System Fuzzy Logic Controller and Optimization PSO Controller to analysis all result. Simulation models (Classical PI Controller, Expert System Fuzzy Logic Controller and Optimization PSO Controller) of this step as shown in Figures 9-11 and simulation results as shown in Figure 16: Simulation Response (pu) Of PI Control, Fuzzy_

PI Control & PSO_PI Control. The results obtained from the PSO-PI based PES are not only superior in the rise time, settling time and overshoot but can prevent from voltage and has improved power quality.

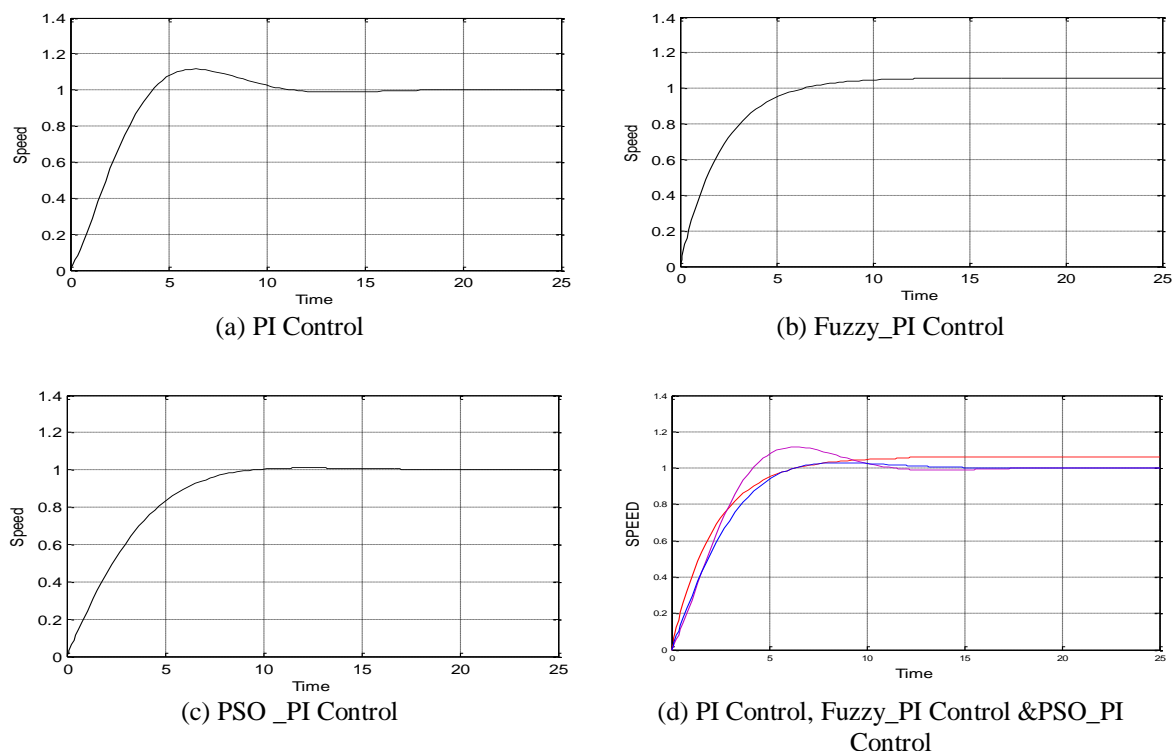


Figure 16. Simulation Response (pu) Of PI Control, Fuzzy_PI Control & PSO_PI Control

6. CONCLUSION

To use different control systems as a case studies. To achieve this objective which characterizes each part of a system such as a DC Motor module, controller and Optimization. After that to investigate the design connection topology for all components of a DC Motor system in order to study the operation of the system for different environmental conditions. The simulation circuits for DC Motor controllers include all realistic components of the system. These results also confirmed that the maximum permissible value. Modeling, analysis, testing and simulation a DC Motor under different conditions using MATLAB. The performance of DC Motor system are controlled and Optimization by PI, fuzzy and PSO.

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